

Analysis of a UWB Planar Antenna with Split Ring Resonator

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Abstract— In this work a new configuration of a ultrawideband (UWB) microstrip antenna is analyzed. A Split Ring Resonator (SRR) is inserted into a circular opening on the opposite side of the patch. The SRR structure will work as a band stop filter. Three configurations are discussed and one of these configurations has a bandwidth from 3.2 to 10.6 GHz and notching a frequency band between 5.3 to 6.0 GHz. The prototype is built and results measured and simulated are compared.

Index Terms— Microstrip Antenna, SRR, UWB.

I. INTRODUCTION

The ultra-wideband communication system (UWB) occupies the frequency range 3.1 to 10.6 GHz, which is approved by the Federal Communications Commission [1]. UWB technology is a short-range wireless technology for transmitting large amounts of data and high speeds with very low power. To establish communication between two nodes, the UWB transceivers require antennas, preferably small in size and low production cost [2]. Several slit antennas for UWB application have been reported so far [3-5]. To avoid interference between UWB system and the system (WLAN), wireless local area network, a filter rejects-band on UWB system is needed. However, the use of a filter will increase the complexity of the UWB system. Therefore, a UWB antenna having reject-band for certain characteristic frequencies is an alternative to overcome this problem. Multiple antennas to reject-band have been reported [6-10].

In [10] the SRR is used in a truncated ground plane, obtaining a bandwidth of 3.6 to 9.2 GHz with a reject band that covers the frequency of 5.5GHz. This work propose a new micro strip antenna UWB with patch circular with rejects-

band designed for a resonant frequency of 5.8 GHz. The cutting band is achieved by inserting a conventional split ring resonator (SRR) in the circular opening made in the ground plane, which is magnetically excited by means of a circular patch on the opposite side of the dielectric substrate.

The SRR is composed of two strips of concentric rings with an opening. The two rings are positioned such that the opening of each ring is placed in front of the other. This opening tightly controls the resonant frequency. Basically, the SRR behaves as an LC resonator and the resonance frequency can be calculated as follows [11],

$$\omega_0 = \sqrt{\frac{2}{\pi r_0 L_0 C}}, \quad (1)$$

where L_0 is the inductance per unit length of the rings, C is the total capacitance of the SRR, and r_0 is the mean radius of the two rings.

The SRR may be magnetically excited, if the rings are oriented correctly. To properly excite the SRR by a varying magnetic field, a significant component in the axial direction is required.

II. DESIGN OF THE ANTENNA

The geometry of the antenna pattern is a planar structure with a circular patch with a radius $r = 6.0$ mm which is connected to the supply line with a width $w = 3.0$ mm and a length $l = 4.0$ mm. The antenna is built on an FR4 substrate 30 (a) x 30 (b) mm² and thickness 1.56 mm, dielectric constant $\epsilon_r = 4.4$ and loss tangent $\tan \delta = 0.02$. A conductive plane is placed on one side of the substrate, which is the ground plane. For the proposed new antenna, it is considered the antenna parallel to the xy plane and centered at the origin of a Cartesian coordinate system (x, y, z). In this case, a circular opening centered on the origin of radius $R = 11.0$ mm is made in the ground plane. The system center is aligned along the y-axis. To implement the property of rejects band, the SRR is inserted into the circular opening in the ground plane underneath the patch radiating, as shown in Figure 1. Figure 1 also shows the standard antenna. This arrangement can achieve high magnetic coupling between the patch and the rings at resonance. The presence of the rings leads to a negative effective permeability within a narrow band above the resonance, where the signal propagation is inhibited.

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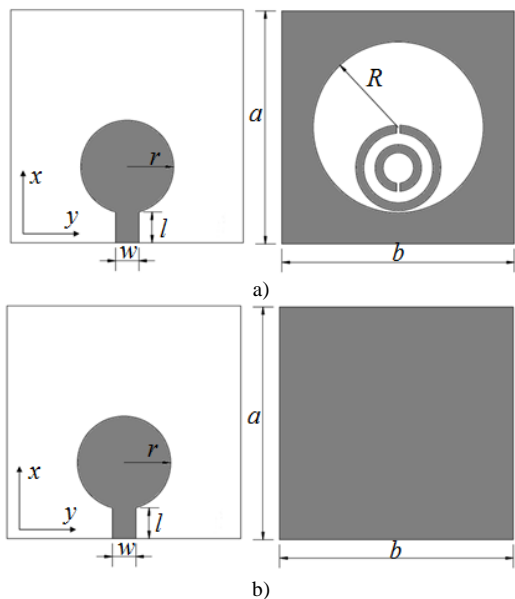


FIG. 1. Geometry of the proposed(a) and standard(b) antenna.

Figure 2 shows the geometry of the proposed SRR. The SRR element for a resonant frequency of 5.5 GHz dimensions are obtained as follows: $r_1 = 5.5$ mm, $r_2 = 3.0$ mm, $S = 1.0$ mm, $D = 1.5$ mm and $G = 0.5$ mm [10].

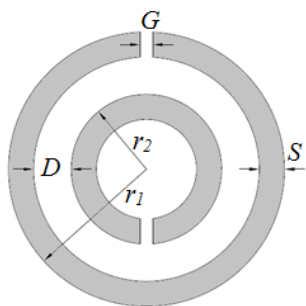


FIG. 2. Geometry of the proposed SRR.

To optimize the position of the SRR, three different configurations were tested (Figure 3): one below the patch (10, 15), a second positioned in the center of the ground plane (15, 15) and a third above the patch (20, 15). These antennas are called antenna 1, 2 and 3, respectively.

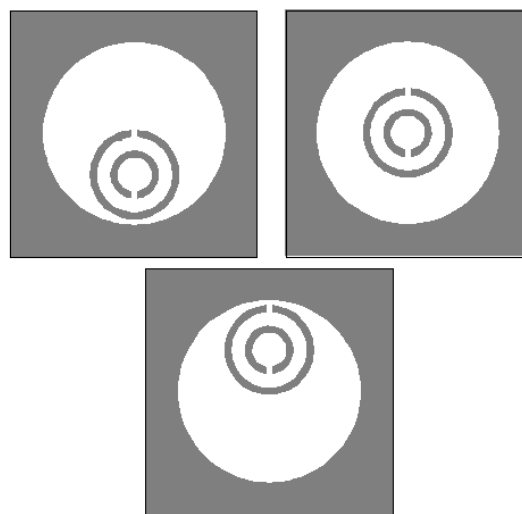


FIG. 3. Geometry of the ground plane of the antenna 1, 2 and 3.

After modeling the standard antenna and the proposed settings, simulations were performed in order to obtain comparative data. A prototype was built for the antenna configuration 1 and standard antenna. The results of the simulations and the measured results of the prototypes constructed are described in the next section.

III. NUMERICAL INVESTIGATIONS AND EXPERIMENTAL RESULTS

For the standard antenna and the proposed settings described in the previous section, simulations to obtain values of S_{11} versus frequency were performed. All these values were compared in order to verify the characteristics of all the settings.

Figure 4 shows the simulation of S_{11} as function of the frequency for the standard antenna and proposed settings using the SRR.

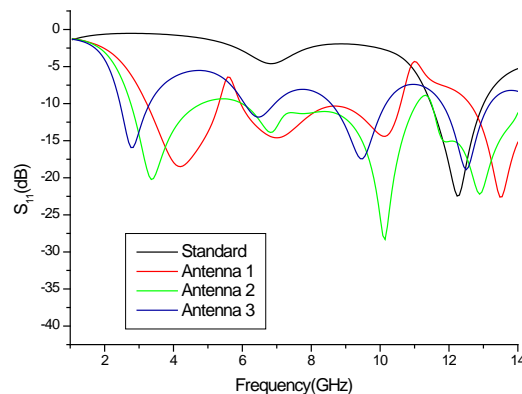


FIG. 4. Values of S_{11} as function of the frequency to standard antenna and proposed settings.

It is can see that with the inclusion of this ground plane modified by inserting the SRR, the antenna works well in some frequencies in the range 2-13 GHz, which does not occur with the standard antenna. The band of rejection arises due to the resonator and the antenna 1 has better bandwidth with the desired rejection band, while that the antenna 2 does not create the rejection band and the antenna 3 has a lower bandwidth. The bandwidth of this antenna 1 is of 3.2 to 10.6 GHz with a rejection band of 5.3 to 6.0 GHz, which makes it a candidate antenna for UWB communication systems.

The gain was also analyzed and Figure 5 shows the radiation patterns in 2D and 3D on the E-plane ($\phi = 0^\circ$) and H-plane ($\phi = 90^\circ$), where the antenna proposal has omnidirectional radiation pattern in the H-plane (xz plane) and almost omnidirectional in the E-plane (yz plane), for all configurations.

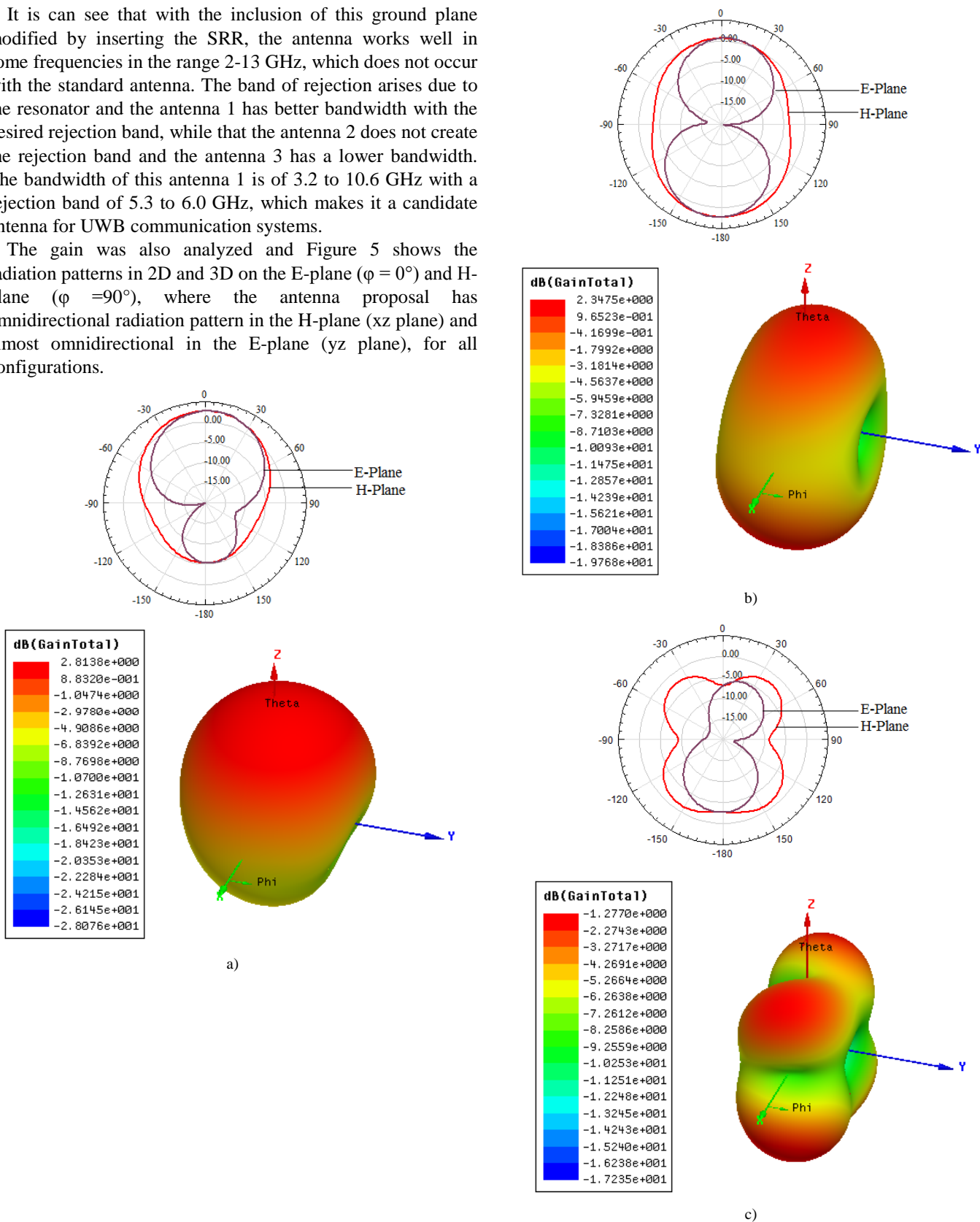


FIG. 5. Simulated results of radiation patterns in 2D and 3D of the E-plane and H-plane to 5.8 GHz. a) antenna 1, b) antenna 2, c) antenna 3.

While the antenna 1 presents gain of 2.814 dB, the antenna 2 presents 2.348 dB and the antenna 3 presents -1.277 dB.

The prototypes for the standard antenna and the antenna 1 are constructed as shown in Figures 6 and 7, respectively.

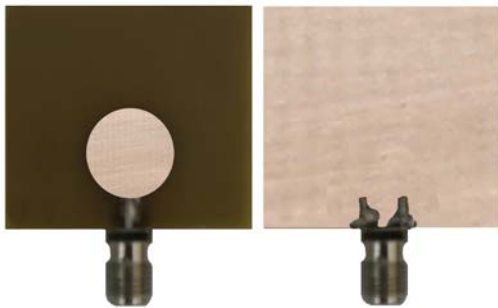


FIG. 6. Standard antenna built.



FIG. 7. Antenna 1 built.

Measurements S_{11} as function of the frequency were performed in Rohde & Schwarz ZVB14, which is a vector Network Analyzer, device that allows the measurement and testing of antennas.

Measurements of S_{11} were performed, and the data from the two antennas, measured and simulated, were compared. This comparison is shown in Figures 8 and 9, respectively.

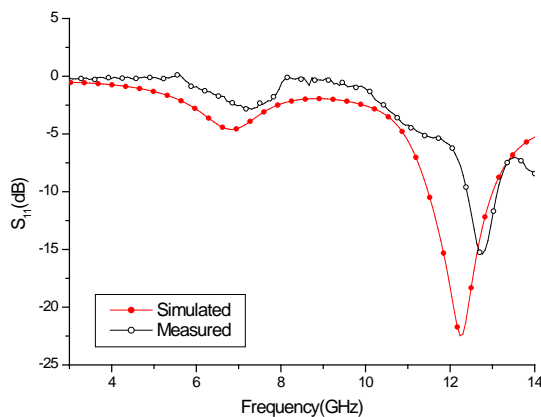


FIG. 8. S_{11} of the standard antenna.

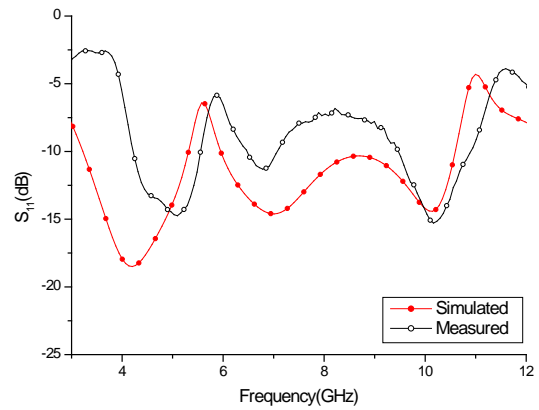


FIG. 9. S_{11} of the antenna 1.

This comparison shows a good agreement between measured and simulated values. Differences between the two curves are likely to give the misalignment between the two sides during the manufacturing process, and inaccuracies in dimensions.

IV. CONCLUSIONS

In this paper, a micro strip antenna with Split Ring Resonator - SRR was analyzed using computer simulation. Three configurations are analyzed by varying the position of the SRR. The proposed structure presented an excellent bandwidth of 3.2 to 10.6 GHz, when calculated as return loss less than -10 dB. A rejection band of 5.3 to 6.0 GHz is obtained using SRR that resonates at the required cut-off frequency. Since FR4 is well-known to present high losses at microwave we can observe in radiation patterns that antennas based on metamaterials have a solution interest to these demands, because its peculiar features not usually found in nature, give rise to a whole new range of electromagnetic applications, in order to increase the efficiency of the structures, in a way that was not possible with classical structures. It was observed that the structure of the proposed antenna becomes a candidate for use in wireless communications systems. Prototypes were built and measured data were compared with the simulated.

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