THE HARTLEY TRANSFORM IN A FINITE FIELD

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Abstract - In this paper, the k-trigonometric functions over the Galois Field GF(q) are introduced and their main properties derived. This leads to the definition of the cas_k(.) function over GF(q), which in turn leads to a finite field Hartley Transform. The main properties of this new discrete transform are presented and areas for possible applications are mentioned.

Keywords - Finite field trigonometry, Galois fields, Hartley transform.

1. INTRODUCTION

Discrete transforms play a very important role in engineering. A significant example is the well known Discrete Fourier Transform (DFT), which has found many applications in several areas, specially in electrical engineering. A DFT for finite fields was introduced by Pollard in 1971 [1] and applied as a tool to perform discrete convolutions using integer arithmetic. Since then several new applications of the Finite Field Fourier Transform (FFFT) have been found, not only in the fields of digital signal and image processing [2-5], but also in different contexts such as error control coding and cryptography [6-8].

A second relevant example concerns the Discrete Hartley Transform (DHT) [9], the discrete version of the integral transform introduced by R. V. L. Hartley in [10]. Although seen initially as a tool with applications only on the numerical side and having connections to the physical world only via the Fourier transform, the DHT has proven over the years to be a very useful instrument with many interesting applications [11-13].

In this paper the DHT over a finite field is introduced. In order to obtain a transform that holds some resemblance with the DHT, it is firstly necessary to establish the equivalent of the cosine and sine functions over a finite structure. Thus, in the next section, the k-trigonometric functions cos_k and sin_k are defined from which the cas_k (cosine and sine) function is obtained and used, in section 3, to introduce a symmetrical discrete transform pair, the finite field Hartley transform, or FFHT for short. A number of properties of the FFHT is presented, including the cyclic convolution property and Parseval's relation. In section 4, the condition for valid spectra, similar to the conjugacy constraints for the Finite Field Fourier Transform, is given. Section 5 contains a few concluding remarks and some possible areas of applications for the ideas introduced in the paper. The FFHT presented here is different from an earlier proposed Hartley Transform in finite fields [14] and appears to be the more natural one.

2. K-TRIGONOMETRIC FUNCTIONS

The set G(q) of gaussian integers over the Galois field GF(q) defined below plays an important role in the ideas introduced in this paper (hereafter the symbol := denotes equal by definition).

Definition 1. G(q) := \{a + jb, a, b \in GF(q), q = p^r, r \in \mathbb{N}, j^2 = -1 \} is a quadratic non-residue in GF(q), is the set of gaussian integers over GF(q).

Let \( \otimes \) denote the cartesian product. It can be shown, as indicated below, that the set G(q) together with the operations \( \otimes \) and \( * \) defined below, is a field.

Proposition 1 : Let

\[
\begin{align*}
\oplus & : G(q) \times G(q) \rightarrow G(q) \\
(a_1 + jb_1, a_2 + jb_2) \rightarrow (a_1 + jb_1) \oplus (a_2 + jb_2) = \\
& = (a_1 + a_2) + j(b_1 + b_2)
\end{align*}
\]

and

\[
\begin{align*}
* & : G(q) \times G(q) \rightarrow G(q) \\
(a_1 + jb_1, a_2 + jb_2) \rightarrow (a_1 + jb_1) * (a_2 + jb_2) = \\
& = (a_1 a_2 - b_1 b_2) + j(a_1 b_2 + a_2 b_1).
\end{align*}
\]

The structure \( G(q) := < G(q), \otimes, * > \) is a field. In fact, \( G(q) \) is isomorphic to \( GF(q^2) \).

Trigonometric functions over the elements of a Galois field can be defined as follows.

Definition 2. Let \( a \) have multiplicative order \( N \) in \( GF(q), q = p^r, p \neq 2 \). The \( GI(q) \) valued k-trigonometric functions of
\( \angle(\alpha^t) \) in GF(q) (by analogy, the trigonometric functions of \( k \) times the "angle" of the "complex exponential" \( \alpha^t \)) are defined as

\[
\cos_k(\angle \alpha^t) := \frac{1}{2}(\alpha^{kt} + \alpha^{-kt})
\]
and

\[
\sin_k(\angle \alpha^t) := \frac{1}{2}(\alpha^{kt} - \alpha^{-kt}),
\]
for \( i, k = 0, 1, \ldots, N-1 \).

For simplicity suppose \( \alpha \) to be fixed. We write

\[
\cos_k(\alpha^i) \quad \text{as} \quad \cos_k(i)
\]
and

\[
\sin_k(\alpha^i) \quad \text{as} \quad \sin_k(i).
\]
The \( k \)-trigonometric functions satisfy properties P1-P11 below.

**P1. Unit Circle:** \( \sin^2(\alpha^i) + \cos^2(\alpha^i) = 1 \).

**Proof:**

\[
\sin^2(\alpha^i) + \cos^2(\alpha^i) = \frac{1}{2}(\alpha^{2ki} - \alpha^{-2ki}) + \frac{1}{2}(\alpha^{2ki} + \alpha^{-2ki}) = 1
\]

**P2. Even / Odd:** \( \cos_k(i) = \cos_k(-i) \), \( \sin_k(i) = -\sin_k(-i) \).

**Proof:**

\[
\cos_k(-i) = \frac{1}{2}(\alpha^{-ki} + \alpha^{ki}) = \cos_k(i); \quad \sin_k(-i) = \frac{1}{2}(\alpha^{-ki} - \alpha^{ki}) = -\sin_k(i).
\]

**P3. Euler Formula:** \( \alpha^{kt} = \cos_k(i) + jsin_k(i) \).

**Proof:**

\[
\cos_k(i) + jsin_k(i) = \frac{1}{2}(\alpha^{ki} + \alpha^{-ki}) + \frac{j}{2}(\alpha^{ki} - \alpha^{-ki}) = \alpha^{kt}
\]

**P4. Addition of Arcs:**

\[
\cos_k(i + t) = \cos_k(i)\cos_k(t) - \sin_k(i)\sin_k(t), \quad \sin_k(i + t) = \sin_k(i)\cos_k(t) + \cos_k(i)\sin_k(t).
\]

**Proof:**

\[
\cos_k(i + t) = \frac{1}{2}(\alpha^{ki} + \alpha^{kt} + \alpha^{-ki} + \alpha^{-kt}) = \frac{1}{2}(\cos_k(i) + \cos_k(t)) + \frac{1}{2}(\sin_k(i) - \sin_k(t)).
\]
The proof for the \( \sin(.) \) function is similar.

**P5. Double arc:**

\[
\cos_k^2(i) = \frac{1 + \cos_k(2i)}{2}, \quad \sin_k^2(i) = \frac{1 - \cos_k(2i)}{2}.
\]

**Proof:** According to P4, \( \cos_k(2i) = \cos_k^2(i) - \sin_k^2(i) = \cos_k^2(i) - (1 - \cos_k^2(i)) = 2\cos_k^2(i) - 1 \), and the result follows.
The proof for the \( \sin(.) \) function is similar.

**P6. Symmetry:** \( \cos_k(i) = \cos_k(k) \), \( \sin_k(i) = \sin_k(k) \).

**Proof:** Follows directly from definition 2.

**P7. Complementary Symmetry:** With \( itk \neq 0 \) and \( k+t = i+r = N \),

\[
\cos_k(i) = \cos_k(t), \quad \sin_k(i) = -\sin_k(t).
\]

**Proof:**

\[
2[\cos_k(i) - \cos_k(t)] = (\alpha^{ki} + \alpha^{-ki}) - (\alpha^{kt} + \alpha^{-kt}) = (\alpha^{ki} - \alpha^{kt}) = 0
\]
since \( \alpha \) has order \( N \).

**P8. Periodicity:**

\[
\cos(i + N) = \cos(i), \quad \sin(i + N) = \sin(i).
\]

**Proof:**

\[
\cos(i + N) = \frac{1}{2}(\alpha^{(i+N)k} + \alpha^{-(i+N)k}) = \frac{1}{2}(\alpha^{ik} + \alpha^{-ik}) = \cos(i),
\]

since \( \alpha \) has order \( N \).

**P9. Complement:**

\[
\cos(i) = \cos(i), \quad \sin(i) = -\sin(i).
\]

**Proof:**

\[
2[\cos(i) - \cos(t)] = (\alpha^{ik} + \alpha^{-ik}) - (\alpha^{tk} + \alpha^{-tk}) = (\alpha^{ik} - \alpha^{tk}) = 0
\]

The proof for the \( \sin(.) \) function is similar.

**P10. Cosine Summation:**

\[
\sum_{k=0}^{N-1} \cos_k(i) = \begin{cases} N, & i = 0; \\ 0, & i \neq 0 \end{cases}
\]

**Proof:**

\[
\sum_{k=0}^{N-1} \cos_k(i) = \frac{1}{2}(\alpha^{ik} + \alpha^{-ik}) = \frac{1}{2}(\sum_{k=0}^{N-1} \alpha^{ik} + \sum_{k=0}^{N-1} \alpha^{-ik}) = \frac{1}{2}(\alpha^{iN} - 1).
\]

If \( i = 0 \) then \( \sigma_i = N \). Otherwise \( \sigma_i = \frac{1}{2}(\alpha^{iN} - 1) \).

**P11. Sine Summation:**

\[
\sum_{k=0}^{N-1} \sin_k(i) = 0.
\]

**Proof:**

\[
\sum_{k=0}^{N-1} \sin_k(i) = \frac{1}{2}(\alpha^{iN} - 1).
\]

If \( i = 0 \) then \( \sigma_i = 0 \). Otherwise \( \sigma_i = \frac{1}{2}(\alpha^{iN} - 1) \).
A simple example is given to illustrate the behaviour of such functions.

Example 1. Let $\alpha = 3$, a primitive element of $GF(7)$. The $\cos_k(i)$ and $\sin_k(i)$ functions take the following values in $GF(7)$:

<table>
<thead>
<tr>
<th>$k$</th>
<th>0</th>
<th>1</th>
<th>2</th>
<th>3</th>
<th>4</th>
<th>5</th>
</tr>
</thead>
<tbody>
<tr>
<td>$\cos_k(i)$</td>
<td>1</td>
<td>1</td>
<td>4</td>
<td>3</td>
<td>6</td>
<td>3</td>
</tr>
<tr>
<td>$\sin_k(i)$</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
</tbody>
</table>

The $k$-trigonometric functions have interesting orthogonality properties, such as the one shown in lemma 1.

Lemma 1. The $k$-trigonometric functions $\cos_k(.)$ and $\sin_k(.)$ are orthogonal in the sense that

$$A := \sum_{k=0}^{N-1} [\cos_k(\alpha^i) \sin_k(\alpha^i)] = 0,$$

where $\alpha$ is an element of multiplicative order $N$ in $GF(q)$.

Proof: By definition 2,

$$A = \sum_{k=0}^{N-1} \left[ \frac{1}{2} (\alpha^{k\alpha^i} + \alpha^{k\alpha^i}) \frac{1}{2j} (\alpha^{k\alpha^i} - \alpha^{k\alpha^i}) \right] = \frac{1}{4j} \sum_{k=0}^{N-1} (\alpha^{k(i+0)} - \alpha^{k(i+0)} + \alpha^{k(i-1)} - \alpha^{k(i-1)}).$$

Now, if $i = t$, then $A = (0 + 0 + N - N) / 4j = 0$. If $i = -t$, then $A = (N - N + 0 - 0) / 4j = 0$. Otherwise, $A = (0 + 0 + 0 + 0) / 4j = 0$. 

A general orthogonality condition, which leads to a new Hartley Transform, is now presented via the $\text{cas}_k(\alpha^i)$ function. The notation used here follows closely the original one introduced in [10].

We now introduce the $\text{cas}_k(\alpha^i)$ function which plays an important role in the definition of a finite field Hartley Transform.

Definition 3. Let $\alpha \in GF(q)$, $\alpha \neq 0$. Then

$$\text{cas}_k(\alpha^i) := \cos_k(\alpha^i) + \sin_k(\alpha^i).$$

We use $\text{cas}_k(i)$ to denote $\text{cas}_k(\alpha^i)$. The $\text{cas}_k(i)$ function satisfies properties C1-C7 below:

C1 - $k$-trigonometric Relations

i) $\text{cas}_k(i + t) = \cos_k(i) \text{cas}_k(t) + \sin_k(i) \text{cas}_k(-t)$.

ii) $\text{cas}_k(i - t) = \cos_k(i) \text{cas}_k(-t) + \sin_k(i) \text{cas}_k(t)$.

iii) $\text{cas}_k(i) \text{cas}_k(t) = \text{cos}_k(i - t) + \sin_k(i + t)$.

Proof:

i) By definition $\text{cas}_k(i + t) = \cos_k(i + t) + \sin_k(i + t)$ and from P1 and P2 we may write

$$\text{cas}_k(i + t) = \cos_k(i) \cos_k(t) + \sin_k(i) \sin_k(t) + \sin_k(i) \cos_k(t) + \cos_k(i) \sin_k(t).$$

ii) The proof is similar to i) above.

iii) $\text{cas}_k(i) \text{cas}_k(t) = \cos_k(i) \cos_k(t) + \sin_k(i) \sin_k(t) + \sin_k(i) \cos_k(t) + \cos_k(i) \sin_k(t),$

and the result follows from P2 and P4.

C2. Symmetry: $\text{cas}_k(i) = \text{cas}_k(k - i)$

Proof: Follows directly from P6.

C3. Quadratic Norm

$$[\text{cas}_k(i)]^2 = \cos_k(2i) = \cos_k(i) + \sin_k(i).$$

Proof: With $\text{cas}_k(i) = a + jb$, then

$$(\text{cas}_k(i))^2 = a^2 + j^2 b^2 = a - jb.$$

Therefore

$$[\text{cas}_k(i)]^2 = \text{cos}_k(i) + [\text{sin}_k(i)]^2.$$

C4. Primitivity: If $\text{cas}_k(i)$ is primitive in $GI(q)$ then $\text{cas}_k(i)$ is primitive in $GF(q)$.

Proof: If $\text{cas}_k(i)$ is primitive in $GI(q)$ then the least integer $n$ such that $(\text{cas}_k(i))^n = 1$ is $n = (q+1)(q-1)$. From C4

$$[\text{cas}_k(i)]^{q+1} = (\text{cos}_k(i))^2 + (\text{sin}_k(i))^2 = 1.$$

and the least integer $r$ such that $(\text{cas}_k(i))^r = 1$ is $m = q-1$. Since $[\text{cas}_k(i)]^2$ is in $GF(q)$, that implies it is
primitive.

C5. Periodicity: \( \text{cas}_k(i+N) = \text{cas}_k(i) \).

Proof: It follows directly from the fact that both \( \cos_k(i) \) and \( \sin_k(i) \) functions are periodic with period \( N \).

C6. Nulity: \( \text{cas}_k(i) = 0, \forall i,k = 0, 1, ..., N-1 \).

Proof: The \( \text{cas}_k(i) \) function can be written as
\[
\text{cas}_k(i) = \frac{1}{2} (\alpha^i + \alpha^{-i}) + \frac{1}{2j} (\alpha^i - \alpha^{-i}) = \\
\frac{1}{2} [(\alpha^i + j\alpha^{-i}) - j(\alpha^i + j\alpha^{-i})],
\]
so that
\[
\text{cas}_k(i) = \frac{1 - j}{2} (\alpha^i + j\alpha^{-i}).
\]
Supposing now that \( \text{cas}_k(i) = 0 \), we have
\[
\alpha^i + j\alpha^{-i} = 0, \text{ i.e., } \alpha^{2i} = -j,
\]
which is an absurd since \( j \notin GF(q) \).

C7. Conjugacy Relations
i) \( \text{cas}_k(i) \) = \[\sec_k(2i) + \tan_k(2i)][\text{cas}_k(i)]^\approx \).
ii) \( [\text{cas}_k(i)]^2 + [\text{cas}_k(i)]^\approx = 2 \), where \( ^\approx \) denotes complex conjugate.

Proof: i) From C3 and C7 we have, respectively,
\[
[\text{cas}_k(i)]^\approx = \cos_k(2i)
\]
and
\[
[\text{cas}_k(i)]^2 = 1 + \sin_k(2i).
\]
Therefore
\[
[\text{cas}_k(i)]^\approx = \frac{\cos_k(2i)}{1 + \sin_k(2i)},
\]
so that we may write
\[
\text{cas}_k(i) = \frac{1 + \sin_k(2i)}{\cos_k(2i)} [\text{cas}_k(i)]^\approx
\]
and, from C3, the result follows.

ii) As seen above
\[
[\text{cas}_k(i)]^2 = 1 + \sin_k(2i).
\]
Since \( \alpha \in GF(q) \), we may write
\[
[\text{cas}_k(i)]^\approx = 1 - \sin_k(2i),
\]
and the result follows.

The set \( \{\text{cas}_k(i)\}_{k=0, 1, ..., N-1} \), can be viewed as a set of sequences that satisfy the following orthogonality property:

Theorem 1.

\[
H := \sum_{k=0}^{N-1} \text{cas}_k(\alpha^t) \text{cas}_k(\alpha^t) = \begin{cases} N, & i = t \\ 0, & i \neq t \end{cases},
\]
where \( \alpha \) has multiplicative order \( N \).

Proof: From definition 3 it follows that
\[
H = \sum_{k=0}^{N-1} [\cos_k(i) \cos_k(t) + \sin_k(i) \sin_k(t)] + [\sin_k(i) \cos_k(t) + \cos_k(i) \sin_k(t)],
\]
which, by lemma 1, is the same as
\[
H = \sum_{k=0}^{N-1} \cos_k(i) \cos_k(t) + \sin_k(i) \sin_k(t).
\]
It follows then, from property P4, that
\[
H = \sum_{k=0}^{N-1} \cos_k(i-t),
\]
and, from P9, the result follows.

3. THE FINITE FIELD HARTLEY TRANSFORM

Definition 4. Let \( v = (v_0, v_1, ..., v_{N-1}) \) be a vector of length \( N \) with components over \( GF(q) \), \( q = p^r \), \( p \neq 2 \). The Finite Field Hartley Transform (FFHT) of \( v \) is the vector \( V = (V_0, V_1, ..., V_{N-1}) \) of components \( \forall i \in GF(q^r) \), given by
\[
V_k := \sum_{i=0}^{N-1} v_i \text{cas}_k(\alpha^i),
\]
where \( \alpha \) is a specified element of multiplicative order \( N \) in \( GF(q^r) \).

Such a definition clearly mimics the classical definition of the Discrete Hartley Transform [9]. The inverse FFHT is given by the following theorem.

Theorem 2. The \( N \)-dimensional vector \( v \) can be recovered from its Hartley discrete spectrum \( V \) according to
\[
v_i = \frac{1}{N (\text{mod} \ p)} \sum_{k=0}^{N-1} V_k \text{cas}_k(\alpha^t).
\]

Proof: Let \( \tilde{v}_i := \frac{1}{N (\text{mod} \ p)} \sum_{k=0}^{N-1} V_k \text{cas}_k(\alpha^t) \).

After substituting the \( V_k \) given by definition 4 in the above expression, it follows that
The Hartley Transform in a Finite Field

\[ \hat{v}_i = \frac{1}{N(\text{mod} \, p)} \sum_{k=0}^{N-1} v_k \cdot \text{cas}_k(i) \cdot \text{cas}_k(i) \]

Changing the order of summation,

\[ \frac{1}{N(\text{mod} \, p)} \sum_{i=0}^{N-1} v_i \cdot \sum_{k=0}^{N-1} \text{cas}_k(i) \cdot \text{cas}_k(i) \]

\[ = \frac{1}{N(\text{mod} \, p)} \sum_{i=0}^{N-1} v_i \begin{cases} 1, & r = i \\ 0, & r \neq i \end{cases} = v_i \]

A signal \( v \) and its discrete Hartley spectrum \( V \) are said to form a finite field Hartley Transform pair, denoted by \( v \leftrightarrow V \) or \( H(v) \). It is worthwhile to mention that the FFHT belongs to a class of discrete transforms for which the kernel of the direct and the inverse transform is exactly the same.

Letting now \( g = \{g_i\} \leftrightarrow G = \{G_k\} \) and \( v = \{v_i\} \leftrightarrow V = \{V_k\} \) denote FFHT pairs of length \( N \), the following set of useful properties can be derived.

H1. Linearity

\[ ag + bv \leftrightarrow aG + bV, \quad \forall a, b \in GF(q). \]

H2. Time Shift

If \( v_i = g_{i-d} \), then \( V_k = \text{cos}_k(d)G_k + \text{sin}_k(d)G_{n-k} \).

Proof:

\[ V_k = \sum_{i=0}^{N-1} g_{i-d} \cdot \text{cas}_k(i). \]

Making \( i - d = r \) leads to

\[ V_k = \sum_{r=-d}^{N-1-d} g_r \cdot \text{cas}_k(d + r). \]

which, from C1 and C5, is the same as

\[ V_k = \sum_{r=0}^{N-1} g_r \cdot \text{cas}_k(r) + \text{sin}_k(d) \cdot \text{cas}_k(-r) \]

and

\[ V_k = \text{cos}_k(d)G_k + \text{sin}_k(d)G_{n-k}. \]

H3. Sum of Sequence (dc term):

\[ V_0 = \sum_{i=0}^{N-1} v_i. \]

H4. Initial Value:

\[ v_0 = \frac{1}{N(\text{mod} \, p)} \sum_{k=0}^{N-1} V_k. \]

H5. Symmetry:

\[ G \leftrightarrow N_G. \]

Proof: Denoting the FFHT of \( G = \{G_k\} \) by \( F = \{F_k\} \), we have

\[ F_k = \sum_{r=0}^{N-1} g_r \cdot \text{cas}_k(r). \]

By the definition of the FFHT this can be written as

\[ F_k = \sum_{r=0}^{N-1} g_r \cdot \text{cas}_r(i) \cdot \text{cas}_k(r). \]

Changing the order of summation,

\[ F_k = \sum_{i=0}^{N-1} g_i \cdot \sum_{r=0}^{N-1} \text{cas}_r(i) \cdot \text{cas}_k(r). \]

Considering the symmetry property of the \( \text{cas}(.) \) function, we may use theorem 1 which implies that the innermost summation is nonzero only when \( i = k \). Therefore we obtain \( F_k = Ng_k \) and the proof is complete.

H6. Time Reversal:

\[ g_i \leftrightarrow G_k. \]

Proof: Denoting \( g_i \leftrightarrow F_k \) we have

\[ F_k = \sum_{i=0}^{N-1} g_i \cdot \text{cas}_k(i). \]

Changing \( -i = r \)

\[ F_k = \sum_{r=0}^{N-1} g_r \cdot \text{cas}_k(-r), \]

which, from C2, is the same as \( G_{N-k} \).

H7. Cyclic Convolution:

If \( \star \) denotes cyclic convolution, then

\[ g \star v \leftrightarrow \text{V} \cdot (GV + GV + GV + GV). \]

where \( G \) and \( V \) denotes, respectively, the sequences \( \{G_{N-k}\} \) and \( \{V_{N-k}\} \).

Proof: The cyclic convolution of \( g \) and \( v \) is given by

\[ g \star v = \sum_{r=0}^{N-1} g_r \cdot v_{i-r}, \]

so that its FFHT is

\[ H(g \star v) = \sum_{i=0}^{N-1} \left[ \sum_{r=0}^{N-1} g_r \cdot v_{i-r} \right] \cdot \text{cas}_k(i). \]

Changing the order of summation and using property H2, we obtain

\[ H(g \star v) = \sum_{r=0}^{N-1} g_r \cdot \left[ \text{cos}_k(r) \cdot V_k + \text{sin}_k(r) \cdot V_{N-k} \right] \]
From definition 3 it is possible to express \( \cos_k(i) \) and \( \sin_k(i) \) in terms of the \( \text{cas}_k(i) \) function. Specifically,

\[
\cos_k(i) = \frac{1}{2}[\text{cas}_k(i) + \text{cas}_k(-i)]
\]

and

\[
\sin_k(i) = \frac{1}{2}[\text{cas}_k(i) - \text{cas}_k(-i)].
\]

Therefore

\[
H(g \star v) = \frac{V_k}{2} \sum_{r=0}^{N-1} g_r[\text{cas}_k(i) + \text{cas}_k(-i)] + \frac{V_{N-k}}{2} \sum_{r=0}^{N-1} g_r[\text{cas}_k(i) - \text{cas}_k(-i)].
\]

or

\[
H(g \star v) = \frac{V_k}{2} (G_k + G_{-k}) + \frac{V_{N-k}}{2} (G_k - G_{-k})
\]

and the result follows.

Theorem 3. The vector \( V = \{V_k\} \), \( V_k \in \mathbb{G}(q^n) \), is the spectrum of a signal \( v = \{v_i\}, v_i \in \mathbb{G}(q) \), \( q = p^r \), if and only if

\[
V_k^q = V_{N-qk}
\]

where indexes are considered modulo \( N, \ i, k = 0, 1, ..., N-1 \) and \( N \mid (q^n - 1) \).

Proof: (\( \Rightarrow \)) From the FFHT definition and considering that \( \mathbb{G}(p^t) \) has characteristic \( p \), it follows that

\[
V_k^q = \left( \sum_{i=0}^{N-1} v_i \text{cas}_k(i) \right)^q = \left( \sum_{i=0}^{N-1} v_i^q \text{cas}_q(i) \right).
\]

If \( v_i \in \mathbb{G}(q) \) \( \forall i \), then \( v_i^q = v_i \). The fact that \( j^2 = -1 \in \mathbb{G}(q) \) if and only if \( q \) is a prime power of the form \( 4s + 3 \), implies that \( j^q = -j \). Hence,

\[
V_k^q = \sum_{i=0}^{N-1} v_i \text{cas}_{N-qk}(i) = V_{N-qk}.
\]

(\( \Leftarrow \)) On the other hand, suppose \( V_k^q = V_{N-qk} \). Then

\[
\sum_{i=0}^{N-1} v_i^q \text{cas}_{N-qk}(i) = \sum_{i=0}^{N-1} v_i \text{cas}_{N-qk}(i).
\]

Now, let \( N-qk = r \). Since \( \gcd(q^n - 1, q) = 1 \), both \( k \) and \( r \) ranges over the same values, which implies

\[
\sum_{i=0}^{N-1} v_i^q \text{cas}_r(i) = \sum_{i=0}^{N-1} v_i \text{cas}_r(i),
\]

\( r = 0, 1, ..., N-1 \). By the uniqueness of the FFHT, \( v_i^q = v_i \) so that \( v_i \in \mathbb{G}(q) \) and the proof is complete.

Example 2. With \( q = p = 3, r = 1, m = 5 \) and \( GF(3^5) \) generated by the primitive polynomial \( f(x) = x^2 + x + 2 \), a FFHT of length \( N = 11 \) may be defined by taking an element of order 11 (\( c^{22} \) is such an element, where \( f(c) = 0 \)). The vectors \( v \) and \( V \) given below are an FFHT pair.

\[
v = (0, 1, 0, 2, 0, 0, 0, 1, 0, 2) \leftrightarrow V = (0, j^{17}, j^{20}, j^{29}, j^{57}, j^{19}, j^{40}, j^{79}, j^{50}, j^{37}, j^{50})
\]

The relation for valid spectra shown above implies that only two components \( V_k \) are necessary to completely specify the vector \( V \), namely \( V_0 \) and \( V_1 \). This can be verified simply by calculating the cyclotomic classes \( C_s \) induced by theorem 3 which, in this case, are \( C_0 = (0) \) and \( C_1 = (1, 8, 9, 6, 4, 10, 3, 2, 5, 7) \).
5. MODULAR ENERGY

**Definition 5.** A signal \( \mathbf{v} = (v_0, v_1, \ldots, v_{N-1}) \) with components \( v_i \in GF(p) \) has a modular energy \( E \) given by
\[
E := \sum_{i=0}^{N-1} v_i^2 \quad (\text{mod } p).
\]

The energy of a signal vector \( \mathbf{v} \) assumes values over \( GF(p) \). Supposing that \( \mathbf{v} \) over \( GF(p) \) has a finite field transform \( \mathbf{V} \) over an extension field \( GF(p^m) \), there exists a Parseval relationship
\[
\sum_{k=0}^{N-1} V_k^2 = (\sum_{k=0}^{N-1} V_k^2) \cdot (\sum_{k=0}^{N-1} V_k^2) = N \cdot (\sum_{k=0}^{N-1} V_k^2) = N \cdot E.
\]

And, although the symbols \( V_k \) are on the extension field, the right-hand sum lies always on the ground field. Aiming to define an energy density, the summation cannot be carried out over arbitrary index sets, because it may not yield a value over \( GF(p) \).

**Proposition 2.** If \( \mathbf{V} \) is the finite field transform (either Fourier or Hartley) of a signal \( \mathbf{v} \), then the sum
\[
\sum_{k \in C_s} V_k^2 \in GF(p), \quad \text{where } C_s \text{ is the cyclotomic coset whose leader is } s.
\]

Therefore,
\[
E = \sum_{i=0}^{N-1} v_i^2 - s \left( \sum_{k \in C_s} V_k^2 \right).
\]

**Proof:**

a) **FFFT** - The conjugacy class constraint for the Finite Field Fourier Transform states that the inverse transform components are in \( GF(p) \) if and only if \( V_k^p = V_{kp} \), with indexes taken modulo \( N \) [6]. Considering the cyclotomic coset whose leader is \( s \), i.e.,
\[
C_s = \{ s, ps, p^2s, \ldots, p^{m-1}s \}, \quad \text{where } m \text{ is such that } p^m s = s \text{ (mod } p^m - 1) \text{,}
\]
it is possible to write
\[
\sum_{k \in C_s} V_k^2 = V_s^2 + V_{ps}^2 + \ldots + V_{p^{m-2}s}^2 = V_s^2 + V_{ps}^2 + \ldots + V_{p^{m-1}s}^2 = \text{Tr}(V_s^2),
\]

where \( \text{Tr}(x) \) is the trace function [6]. Since \( \text{Tr}(x) \in GF(p) \) for every \( x \in GF(p^m) \), the case for the FFFT is settled.

b) **FFHT** - From theorem 3, the corresponding expression for the Hartley case is
\[
\sum_{k \in H_s} V_k^2 = V_s^2 + V_{ps}^2 + V_{p^2s}^2 + \ldots + V_{p^{m-1}s}^2 = \sum_{i=0}^{m-1} (-1)^i V_s^2.
\]

where \( H_s \) denotes the Hartley cyclotomic coset whose leader is \( s \). Since \( p \) and \( m \) are both odd, it will happen that \( p^{ms} s = -s \) (mod \( p^m - 1) \) and a repetition will occur only at \( p^{2ms} \), which means that the Hartley cosets are given by the union of the Fourier cosets of \( s \) and \(-s\), i.e.,
\[
H_s = F_s \cup F_{-s},
\]

where \(-s\) is the reciprocal modulo \( N \) of \( s \). Therefore
\[
\sum_{k \in H_s} V_k^2 = \text{Tr}(V_s^2) + \text{Tr}(V_{-s}^2),
\]

which belongs to \( GF(p) \).

**Definition 6.** The cyclotomic density of modular energy (spectral density over a finite field) is defined by
\[
G_s = \sum_{k \in C_s} V_k^2.
\]

Indeed, \( \sum_s G_s = E \), which is similar to \( \int_{-\infty}^{\infty} G(f) df = E \) in the continuous case. Since the sum is carried out over all the frequencies, we can interpret the cyclotomic sets as some sort of "cyclotomic frequencies". It is interesting to consider the particular case in which we have no extension on the alphabet, that is, when the transform is defined from \( GF(p) \) to \( GF(p) \), or even from \( GF(p) \) to \( GF(p) \). In such a case, each cyclotomic class has two elements: \( k \) and \(-k\). Thus, \( G_k = V_k^2 + V_{-k}^2 \) which corresponds to the frequency \( f \) and \(-f\).

We propose then a spectral representation (finite field spectrum) by plotting the cyclotomic energy density versus cyclotomic frequencies. Each spectral line furnishes its contribution to the total modular energy in the corresponding "cyclotomic frequency".

**c-Spectrograms.** We can also define a finite field spectrogram for such a modular energy spectral density, where each element of \( GF(p) \) corresponds to a color of a \( p \)-color "rainbow". This representation is termed a cyclotomic-spectrogram or a c-spectrogram for short. The (infinite) signal sequence over \( GF(p) \) is partitioned into \( N \)-symbol-blocks and the FFHT is evaluated for each block yielding a spectral representation which is plotted in a 2-dimensional grid (block x cyclotomic frequencies).

**Example 3.** We consider FFHT pairs from \( GF(7) \) to \( GF(7) \), with \( N=6 \) and \( a=3 \) an element of order \( N \) in the multiplicative group of \( GF(7) \).

\[
\begin{align*}
100120 & \leftrightarrow \begin{bmatrix} 4 & 6+5j & 1+2j & 2 & 1+5j & 6+2j \\
023065 & \leftrightarrow \begin{bmatrix} 2 & 6+2j & 6 & 6+6j \\
300212 & \leftrightarrow \begin{bmatrix} 1 & 5+4j & 6j & 0 & j & 5+3j \\
110235 & \leftrightarrow \begin{bmatrix} 5 & 4 & 2+6j & 3 & 2+4j & 4 \\
601024 & \leftrightarrow \begin{bmatrix} 6 & 3+2j & 6+4j & 5 & 6+3j & 3+5j
\end{bmatrix}
\end{align*}
\]
Applying theorem 3 to this case leads to $V_k^3 = V_{b-7k}$, so that
$V_k^3 = V_0$ (k=0), $V_k^3 = V_4$ (k=1), $V_k^3 = V_2$ (k=2) and
$V_k^3 = V_{b-7k}$ (k=3). Observe that the components $V_0$ and $V_3$
are not complex, i.e., they are in the ground field. Therefore
the cyclotomic classes $C_k$ induced by theorem 3 are

$$
C_0 = (0), C_1 = (1, 5), C_2 = (2, 4), C_3 = (3).
$$

Table 2 shows, for each transform pair, the energy
contained in each cyclotomic class

### Table 2. Modular energy in the cyclotomic classes of example 3.

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</table>

Figure 1 below shows the c-spectrogram for the block
(N=6) data sequence over GF(7) (with $\alpha=3$) given in
example 3.

![Figure 1](image1.png)

Figure 1. (a) c-spectrogram of the data sequence in example 3. (b) Energy levels.

5. CONCLUSIONS

In this paper, trigonometry for finite fields was
introduced. In particular, the k-trigonometric functions of
the angle of the complex exponential $\alpha^k$ were defined and
their basic properties derived. From the $\cos_k(\omega\alpha^k)$ and
$\sin_k(\omega\alpha^k)$ functions, the $\cos_k(\omega\alpha^k)$ function was defined and
used to introduce a new Hartley Transform, the Finite Field
Hartley Transform (FFHT). It has been shown that the
cyclotomic coset partition induced by the FFHT is such that
an element and its reciprocal modulo N belongs to the same
class, which implies that the number of FFHT components
that need to be computed to completely specify the spectrum
V is approximately half of the number needed for the Finite
Field Fourier Transform.

The concepts of modular energy and c-spectrograms
were also introduced and the notion of a modular energy
spectral density assuming values on a finite field was
defined.

The FFHT seems to have interesting applications in a
number of areas. Specifically, its use in digital signal
processing, along the lines of the so-called number theoretic
transforms (e.g., Mersenne and Fermat number transforms)
should be investigated. In the field of error control codes,
the FFHT might be used to produce a transform domain
description of the field, therefore providing, possibly, an
alternative to the approach introduced in [6]. Digital
multiplexing and spread spectrum are areas that might also
benefit from the new Hartley transform introduced in this
paper. In particular, new schemes of efficient-bandwidth
code-division-multiple-access for band-limited channels
based on the FFHT are currently under development.

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The Hartley Transform in a Finite Field


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